# INTEGRATED CIRCUITS



Product data Supersedes data of 2002 Mar 25 2003 Nov 12



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# SA58605

#### DESCRIPTION

The SA58605 incorporates two op amps and 2.5 V shunt regulator in an unique circuit configuration. The output of the device is inverted when the inverting inputs of either or both op amps exceed the internally set reference voltages at their non-inverting inputs. Amp "A" is referenced to 2.5 V while Amp "B" is referenced to 154 mV. The SA58605 incorporates a "NOR logic" configuration with these specific gate levels.

SA58605 supports voltage control and sensor applications such as AC adapter, switch mode power supply and battery chargers. It is available in a 5-lead small outline surface mount package (SOP003).

#### FEATURES

- Low input bias current: 30 nA typ.
- Low operating supply current: 1.2 mA typ.
- Reference voltages at non-inverting inputs:
  - Amp "A" at 2.5 V typ.
  - Amp "B" at 154 mV

#### SIMPLIFIED DEVICE DIAGRAM



Figure 1. Simplified system diagram.



#### **APPLICATIONS**

- AC adapter and battery charger
- Switched Mode Power Supply (SMPS)
- Control voltage/sensor

### SA58605

#### **ORDERING INFORMATION**

	PACKAGE	TEMPERATURE				
	NAME	DESCRIPTION VERSION				
SA58605D	SOT23-5, SOT25, SO5	Plastic small outline package; 5 leads; body width 1.6 mm	SOP003	–40 to +85 °C		

#### Part number marking

The package is marked with a four letter code. The first three letters designate the product. The fourth letter, represented by 'x', is a date tracking code.

Part Number	Marking
SA58605D	AJAx

#### **PIN CONFIGURATION**



Figure 2. Pin configuration.

#### PIN DESCRIPTION

PIN	SYMBOL	DESCRIPTION
1	OUT	Output.
2	GND	Ground.
3	B <sub>IN</sub>	Amp B inverting input. Non-inverting input internally set at 154 mV reference voltage.
4	A <sub>IN</sub>	Amp A inverting input. Non-inverting input internally set at 2.5 V reference voltage.
5	V <sub>CC</sub>	Positive supply.

#### INTERNAL EQUIVALENT CIRCUIT



Figure 3. Internal equivalent circuit.

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#### **MAXIMUM RATINGS**

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
V <sub>CC</sub>	Supply voltage	-0.3	+20	V
T <sub>amb</sub>	Ambient operating temperature	-40	+85	°C
T <sub>stg</sub>	Storage temperature	-40	+125	°C
Р	Power dissipation	-	250	mW

#### **ELECTRICAL CHARACTERISTICS**

 $T_{amb}$  = 25 °C,  $V_{CC}$  = 5 V (see Figure 6 "Test circuit", and Table 1 "Parameter test circuit 1 and power supply settings"), unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I <sub>CC</sub>	Supply current	$A_{IN} = 0 \text{ V}; B_{IN} = 0 \text{ V}; R_{L} = \infty$	-	1.2	1.7	mA
A amplifier						
V <sub>o(A)</sub>	Output inverting voltage (A)	$B_{IN} = 0 \text{ V}; \text{ R}_{L} = 4.3 \text{ k}\Omega$	2.45	2.50	2.55	V
I <sub>i(bias)(A)</sub>	Input bias current (A)	$B_{IN}$ = 0 V; $R_L$ = 4.3 k $\Omega$	-	30	150	nA
PSSR (A)	PSSR (A)	$B_{IN}$ = 0 V; $R_L$ = 4.3 k $\Omega$	62	-	-	dB
I <sub>o(sink)(A)</sub>	Output sink current (A)	$A_{IN} = 2.7 \text{ V}; B_{IN} = 0 \text{ V}; V_{OUT} = 1.5 \text{ V}$	5	-	-	mA
B amplifier						
V <sub>o(B)</sub>	Output inverting voltage (B)	$A_{IN} = 0 \text{ V}; \text{ R}_{L} = 4.3 \text{ k}\Omega$	151	154	157	mV
I <sub>i(bias)(B)</sub>	Input bias current (B)	$A_{IN} = 0 \text{ V}; \text{ R}_{L} = 4.3 \text{ k}\Omega$	-	30	150	nA
PSSR (B)	PSSR (B)	$A_{IN} = 0 \text{ V}; \text{ R}_{L} = 4.3 \text{ k}\Omega$	65	-	-	dB
I <sub>o(sink)(B)</sub>	Output sink current (B)	$A_{IN} = 0 \text{ V}; B_{IN} = 0.17 \text{ V}; V_{OUT} = 1.5 \text{ V}$	5	-	_	mA

#### **TYPICAL PERFORMANCE CURVES**



Figure 4. Output inverting voltage (A) versus V<sub>CC</sub>.



Figure 5. Output inverting voltage (B) versus V<sub>CC</sub>.

### **TEST CIRCUITS**



Figure 6. Test circuit.

Table 1.	Parameter	test circuit 1	switch and	power	supply	settings

SYMBOL	PARAMETER	S1	S2	S3	S4	S5	S6	S7	S8	S9	S10	S11	S12	R <sub>L</sub> (Ω)	V <sub>IN</sub> (V)	Comments
I <sub>CC</sub>	Power supply current	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON	ON	OFF	ON	OFF	4.3 k		
V <sub>o(A)</sub>	Output inverting voltage (A)	OFF	ON	OFF	OFF	A <sub>IN</sub>	OFF	OFF	ON	OFF	ON	ON	OFF	4.3 k		measure TPA voltage
I <sub>i(bias)(A)</sub>	Input bias current (A)	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON	ON	OFF	ON	OFF	4.3 k		measure TPA voltage
I <sub>o(sink)(A)</sub>	Output sink current (A)	OFF	OFF	OFF	OFF	OFF	OFF	ON	ON	OFF	OFF	OFF	ON		2.7	measure output sink current
PSSR (A)	PSSR (A)	OFF	ON	OFF	ON	OFF	OFF	ON	ON	OFF	OFF	ON	OFF	4.3 k	V <sub>o(A)</sub>	Note 1
V <sub>o(B)</sub>	Output inverting voltage (B)	ON	OFF	OFF	OFF	B <sub>IN</sub>	OFF	OFF	OFF	ON	ON	ON	OFF	4.3 k		measure TPB voltage
l <sub>i(bias)(B)</sub>	Input bias current (B)	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON	ON	OFF	ON	OFF	4.3 k		measure TPB voltage
I <sub>o(sink)(B)</sub>	Output sink current (B)	OFF	OFF	OFF	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	ON		0.17	measure output sink current
PSSR (B)	PSSR (B)	ON	OFF	ON	OFF	OFF	ON	OFF	OFF	ON	OFF	ON	OFF	4.3 k	V <sub>o(B)</sub> – 20 mV	Note 2

#### NOTES:

1.  $V_{OUT1}$  is defined by the voltage when  $V_{CC} = 4$  V.  $V_{OUT2}$  is defined by the voltage when  $V_{CC} = 25$  V. PSSR (A) is shown in Equation (1). 2.  $V_{OUT1}$  is defined by the voltage when  $V_{CC} = 4$  V.  $V_{OUT2}$  is defined by the voltage when  $V_{CC} = 25$  V. PSSR (B) is shown in Equation (1).

$$PSSR = 40 + 20 \log \left| \frac{(25 V - 4 V)}{(V_{OUT1} - V_{OUT2})} \right|$$

Equation (1)

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#### **APPLICATION INFORMATION**

The SA58605 may be used for various voltage control applications in which the input threshold voltage exceed 2.5 V for Amp "A" and 154 mV for Amp "B". When either or both input threshold voltage is exceeded the output is pulled LOW. The output is connected to V<sub>CC</sub> with a pull-up resistance.

Figure 7 shows the schematic for a Universal Converter/Charger circuit in which the SA58605 Amp "A" is used to monitor the B+ level of the converter/charger. Amp "B" input is pulled to ground through a 390  $\Omega$  resistor and will remain there. The input to the Amp "A" is maintained at 0.5  $\times$  B+ with the voltage divider of the two 1.5 k $\Omega$  resistors. As B+ drops below 5 V, the input to Amp "A" of the

SA58605 follows and drops below 2.5 V. This causes the output (pin 1) to go HIGH and the LED is turned off. Q1 base drive is reduced and Q2 is increased. Thus, the PW modulation is increased, and B+ is able to satisfy the load requirements of the battery and circuitry. As B+ is increased above 5 V, it causes the input to Amp "A" to increase above 2.5 V. Then, the output goes LOW and the LED is activated which turns on the detection diode. This increases the drive on Q1 and pulls down the base of Q2 reducing it base drive. With Q2 conducting less, the PW modulation is decreased and B+ is reduced. Under quasi-steady state, B+ is maintained at 5 V.



Figure 7. Universal converter/charger application.

#### Components used in Figure 7

D1-D4: 1N5822

Q1-Q4: 2N3904

T1: Cooper Electronic Technologies Part#: CTX22-25348
Primary: 47 YTurns of #29 AWG, Pin 4 = Start, Pin 5 = Finish Secondary: 5 Turns of 0.40 mm, Pins 1 and 2 = Start, Pins 7 and 8 = Finish Gap: Designed for total primary inductance of 1.24 mH Core: TSF-7070
Bobbin: Pins 3 and 6 removed, EE19

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**PACKING METHOD** 



#### Figure 8. Tape and reel packing method.



SA58605

### SA58605

#### **REVISION HISTORY**

Rev	Date	Description
_2	20031112	Product data (9397 750 12324). ECN 853-2334 30334 of 09 September 2003. Supersedes data of 2002 Mar 25 (9397 750 09865).
		Modifications:
		• Change package outline version to SOP003 in Ordering information table and Package outline sections.
_1	20020325	Product data (9397 750 09865). ECN 853-2334 27919 of 25 March 2002.

#### Data sheet status

Level	Data sheet status <sup>[1]</sup>	Product status <sup>[2] [3]</sup>	Definitions
I	Objective data	Development	This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice.
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#### **Contact information**

For additional information please visit http://www.semiconductors.philips.com. Fax:

sales.addresses@www.semiconductors.philips.com

For sales offices addresses send e-mail to:

Fax: +31 40 27 24825

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Date of release: 11-03

9397 750 12324

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