ASSP SWITCHING REGULATOR CONTROLLER

MB3788

DESCRIPTION

The MB3788 is a dual-channel PWM-type switching regulator controller; it incorporates a reference voltage. The MB3788 has a PWM circuit and an output circuit as well as a reference voltage power supply with a voltage accuracy of \pm 1%. The maximum operating frequency is 1 MHz. It is designed for a voltage-drop output switching regulator suitable for a logic power supply or speed control of a DC motor.

The MB3788 is compatible with all master ICs producing triangular waves, saw-tooth waves and sine waves with an amplitude of 1.3 V to 1.9 V.

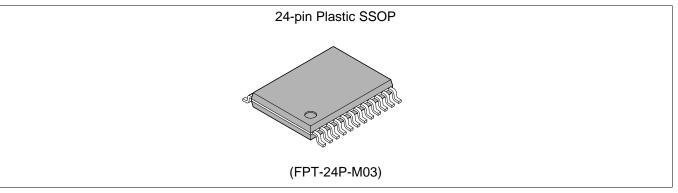
It can be used in high-performance portable equipment such as a video camcorder or notebook personal computer (word processor).

FEATURES

- Wide operating power supply voltage range: 3.6 V to 18 V
- Low power dissipation
- Operating: 1.9 mA (standard) Standby: 10 μA Max

(Continued)

PACKAGE



This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields. However, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum rated voltages to this high impedance circuit.



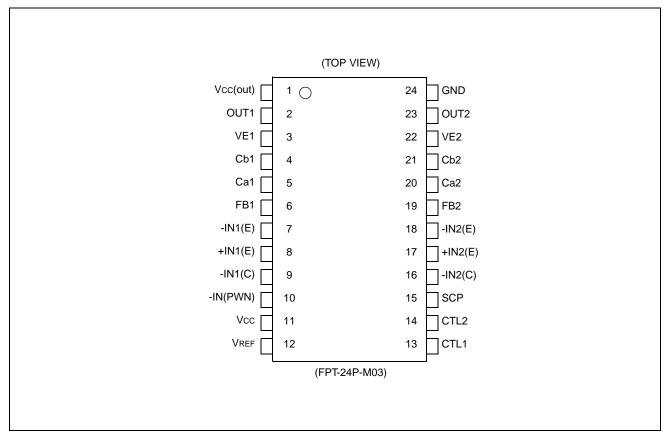
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- High-frequency operation: 100 kHz to 1 MHz
- On-chip timer and latch-type short-circuit detection circuit
- Wide error amplifier input voltage range: -0.2 V to Vcc 1.8 V
- + On-chip high-accuracy reference voltage circuit: 2.50 V $\pm 1\%$
- Output circuit

PNP transistor drive output pin: Push-pull type ON/OFF current values set independently

- On-chip standby function and output control function
- High-density packaging: SSOP-24P



■ PIN ASSIGNMENT

■ PIN DESCRIPTION

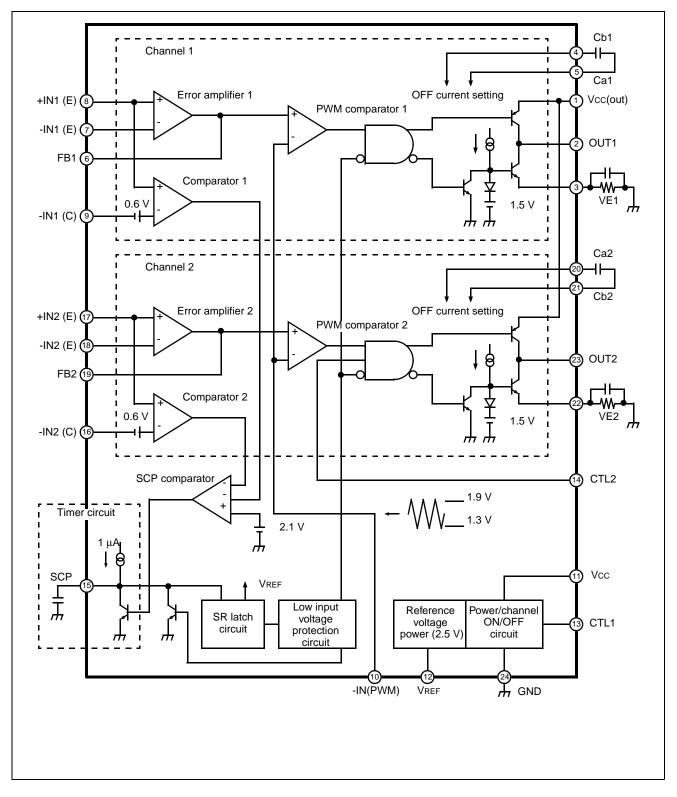
Pin No.		Pin name	I/O	Descriptions
	2	OUT1	0	Channel 1 push-pull type output
	3	VE1	I	Channel 1 output current setting
	4	Ca1	—	Channel 1 output transistor OFF current setting: Output transistor OFF
Channel 1	5	Cb1	—	The current is set by connecting a capacitor between pins Ca1 and Cb1.
	6	FB1	0	Channel 1 error amplifier output
	7	-IN1(E)	I	Channel 1 error amplifier inversion input
	8	+IN1(E)		Channel 1 error amplifier non-inversion input
	9	-IN1(C)	I	Channel 1 comparator inversion input
	16	-IN2(C)	I	Channel 2 comparator inversion input
	17	+IN2(E)	I	Channel 2 error amplifier non-inversion input
	18	-IN2(E)	I	Channel 2 error amplifier inversion input
Channel 2	19	FB2	0	Channel 2 error amplifier output
	20	Ca2		Channel 2 output transistor OFF current setting: Output transistor OFF
	21	Cb2		The current is set by connecting a capacitor between pins Ca2 and Cb2.
	22	VE2	I	Channel 2 output current setting
	23	OUT2	0	Channel 2 push-pull type output
13 CTL1 I H level: Po		I	Power and channel 1 control pin H level: Power and channel 1 operating L level: Standby	
Control circuit	14	CTL2	I	Channel 2 control pin When CTL1 pin = H level, H level: Channel 2 operating L level: Channel 2 OFF
	15	SCP		Short-circuit protection circuit capacitor connection
	1	Vcc ²	—	Output circuit power pin
_	10	-IN(PWM)	I	Master oscillating waveform input
Power circuit	11	VCC1		Reference power and control circuit power
	12	Vref	0	Reference voltage output
	24	GND	—	Ground

Note: The alphabetic characters in parenthesis above indicate the following input pins.

(C): Comparator

(E): Error amplifier

BLOCK DIAGRAM



■ FUNCTIONAL DESCRIPTION

1. Major Functions

(1) Reference voltage power circuit

The reference voltage power supply produces a reference voltage (≈ 2.50 V) which is temperature-compensated by the voltage supplied from the power pin (pin 11); it is used as the IC internal circuit operating power supply. The reference voltage can also be output externally at 1 mA from VREF pin (pin12).

(2) Error amplifier

The error amplifier detects the switching regulator output voltage and outputs a PWM control signal. It has a wide in-phase input voltage range of -0.2 V to Vcc - 1.8 V to make setting from an external power supply easy.

Connecting the output pin and inversion input pin of the error amplifier through a feedback resistor and capacitor allows setting of any loop gain to provide stable phase compensation.

(3) PWM comparator

The PWM comparator controls the output pulse ON time according to the input voltage.

The voltage input to the -IN pin (PWM) turns the output transistor on when it is lower than the output voltage of the error amplifier.

(4) Output circuit

The output circuit is configured in a push-pull form and uses a PNP transistor drive system to drive a transistor of up to 30 mA. (See *How to Set Output Current.*)

2. Channel Control Function

Channels can be set ON/OFF by combining the voltage levels at pin CTL1 (pin 13) and pin CTL2 (pin 14).

Channel ON/OFF Setting Conditions

Voltage lev	el at CTL pin	Channel ON/OFF status			
CTL1 CTL2		Power circuit Channel 1		Channel	
L	×	Stand by state*			
Н	Н	0	ON		
	L	- ON		OFF	

*: The power current in the standby state is 10 μ A Max.

3. Protection Functions

(1) Timer and latch-type short-circuit protection circuit

The SCP comparator detects the output voltage levels of two comparators to detect an output short circuit. If the output voltage of one comparator increases to 2.1 V, the transistor of the timer circuit is turned off and the short circuit protection capacitor connected externally to the SCP pin (pin 15) starts charging.

The latch circuit turns off the output transistor and simultaneously clears the duty cycle to 0 when the output voltage level of the comparator does not return to the normal voltage level until the capacitor voltage rises to the base-emitter junction voltage VBE (≈ 0.65 V) of the transistor. (See *How to Set Time Constant for Timer & Latch-Type Short-Circuit Protection Circuit.*)

When the protection circuit operates, recycle the power to reset the circuit.

(2) Low input voltage malfunction fail-safe circuit

A transient at power-on, or an instantaneous supply voltage drop can cause a control IC malfunction, which may damage the system. The low input voltage malfunction fail-safe circuit detects the internal reference voltage level based on the supply voltage level, resets the latch circuit, turns off the output transistor, clears the duty cycle to 0 and holds the SCP pin (pin 15) at Low level. All circuits are recovered when the supply voltage is greater than the threshold voltage of the fail-safe circuit.

ABSOLUTE MAXIMUM RATINGS

					$(TA = +25^{\circ}C)$
Parameter	Symbol	Conditions	Ratings		Unit
Falameter	Symbol	Conditions	Min	Max	
Supply voltage	Vcc	—	—	20	V
Control input voltage	VICTL	—	—	20	V
Allowable loss	PD	Ta ≤ +25°C	—	500*	mW
Operating ambient temperature	Тор	—	-30	+85	°C
Storage temperature	Tstg	_	-55	+125	°C

*: Value obtained when mounted on 4 cm \times 4 cm double-sided epoxy substrate

WARNING: Semiconductor devices can be permanently damaged by application of stress (voltage, current, temperature, etc.) in excess of absolute maximum ratings. Do not exceed these ratings.

RECOMMENDED OPERATING CONDITIONS

(TA = +25°C)

Parameter	Symbol	Conditions		Unit		
Faranieter	Symbol	Conditions	Min	Тур	Max	Onit
Supply voltage	Vcc	_	3.6	6.0	18	V
Reference voltage output current	IOR	_	-1		0	mA
Error amplifier input voltage	Vi	_	-0.2	_	Vcc - 1.8	V
Error amplifier input voltage	Vi	_	-0.2	_	Vcc	V
Control input voltage	VICTL	_	-0.2	_	18	V
Output current	lo	_	3.0	_	30	mA
Operating frequency	fosc		100	300	1000	kHz
Operating ambient temperature	Тор	_	-30	25	85	°C

WARNING: The recommended operating conditions are required in order to ensure the normal operation of the semiconductor device. All of the device's electrical characteristics are warranted when the device is operated within these ranges.

Always use semiconductor devices within their recommended operating condition ranges. Operation outside these ranges may adversely affect reliability and could result in device failure.

No warranty is made with respect to uses, operating conditions, or combinations not represented on the data sheet. Users considering application outside the listed conditions are advised to contact their FUJITSU representatives beforehand.

■ ELECTICAL CHARACTERISTICS

 $(VCC = 6V, TA = +25^{\circ}C)$

				(VCC = 6V, TA = 4			Unit
	Parameter	Symbol	Conditions	Min Typ Max			
Deferring II		\/					
	Reference voltage	VREF	IOR = -1 mA	2.475	2.500	2.525	V
Reference	Output voltage temperature variation	ΔVref/ Vref	$TA = -30^{\circ} \text{ to } +85^{\circ}\text{C}$	-2	±0.2	2	%
voltage	Input stability	Line	Vcc = 3.6 V to 18 V	—	2	10	mV
	Load stability	Load	IOR = -0.1 mA to 1 mA	—	3	10	mV
	Short-circuit output current	los	Vref = 2 V	-20	-8	-3	mA
	Threshold voltage	VtH	—	—	2.65	—	V
Low voltage malfunction	Threshold voltage	VtL	—	—	2.45	—	V
fail-safe circuit	Hysteresis width	VHYS	—	80	200	—	mV
circuit	Reset voltage	VR	—	1.5	1.9		V
Short-circuit	Input offset voltage	Vio	—	0.58	0.65	0.72	V
detection	Input bias current	Ів	VI = 0 V	-200	-100		nA
comparator	In-phase input voltage range	VICM	_	-0.2	_	Vcc-1.8	V
	Threshold voltage	VtPC	—	0.60	0.65	0.70	V
Short-circuit	Input standby voltage	VSTB	—	—	50	100	mV
detector	Input latch voltage	Vi	—	—	50	100	mV
	Input source current	libpc	—	-1.4	-1.0	-0.6	μA
	Input offset voltage	Vio	VFB = 1.6 V	-10		10	mV
	Input offset current	lio	VFB = 1.6 V	-100	_	100	nA
	Input bias current	Ів	VFB = 1.6 V	-200	-60		nA
	In-phase input voltage range	VICM	—	-0.2		Vcc-1.8	V
	Voltage gain	Av	—	60	100		dB
Error amplifier	Frequency bandwidth	BW	Av = 0 dB	—	800		kHz
	In-phase signal rejection ratio	CMRR	_	60	80		dB
		Vom+	—	Vref-0.3	2.4	—	V
	Maximum output voltage width	Vom-	—		0.05	0.5	V
	Output sink current	IOM+	VFB = 1.6 V	—	120	—	μA
	Output source current	Іом-	VFB = 1.6 V		-2		mA

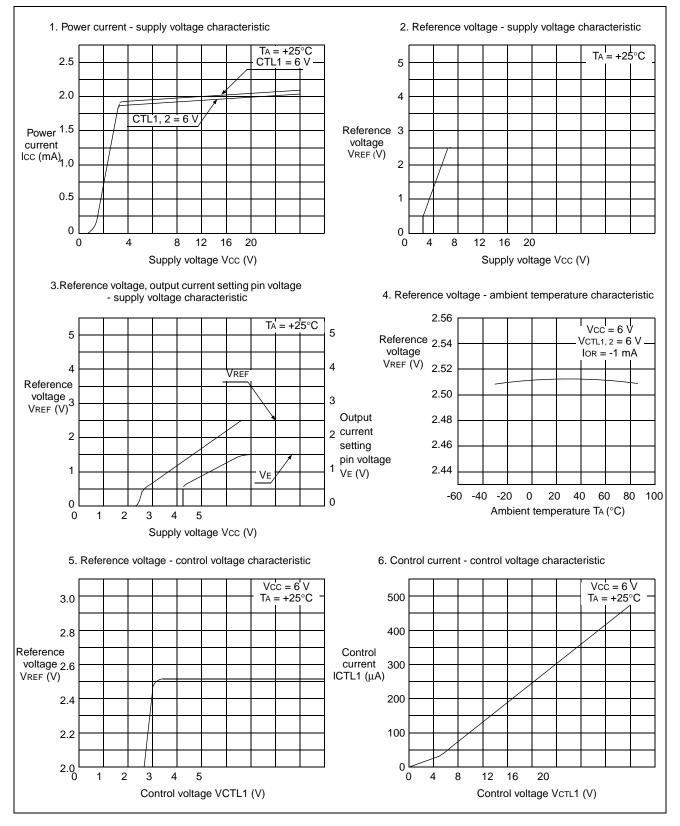
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	Paramotor		Conditions	Values			Unit
Parameter		Symbol	Conditions	Min	Тур	Max	
	Theoretical designs	Vt0	Duty cycle = 0 %	1.05	1.3	—	V
	Threshold voltage	Vt100	Duty cycle = 100 %	—	1.9	2.25	V
PWM comparator	Input sink current	IIM+	—	—	120	—	μA
	Input source current	Ім-	—	—	-2	—	mA
	Input bias current	Ів	VI = 0 V	-1.0	-0.5	-	μA
Control	Threshold voltage	Vth	—	0.7	1.4	2.1	V
	Input current	Ін	VCTL = 5 V	—	100	200	μA
		lı∟	VCTL = 0 V	-10	—	10	μA
	Source current	lo	—	—	-40	—	mA
Output	Sink current	lo	RB = 50 Ω	18	30	42	mA
	Output leak current	Ilo	Vo = 18 V	-	—	20	μA
All devices	Standby current	Icco	—	[—	0	10	μA
	Power current at output OFF	Icc	—	—	1.9	2.7	mA

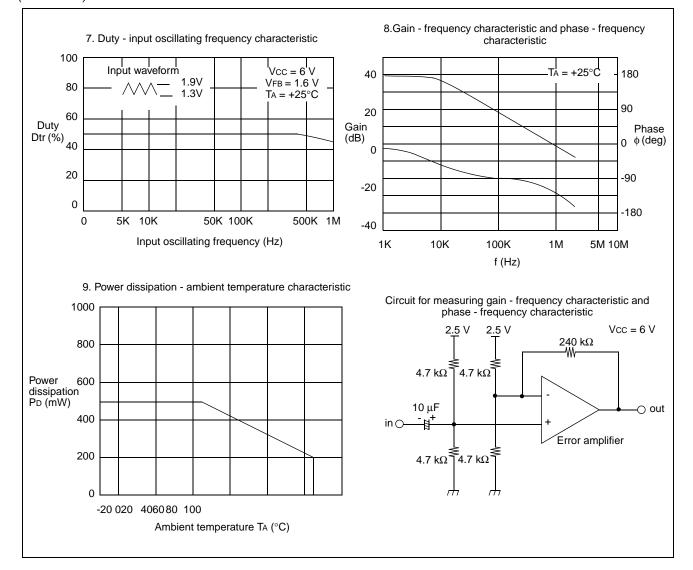
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■ STANDARD CHARACTERISTIC CURVES



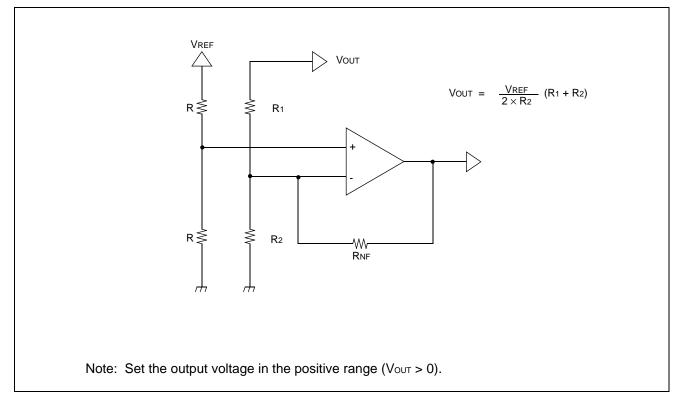
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■ HOW TO SET OUTPUT VOLTAGE



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■ HOW TO SET OUTPUT CURRENT

The output circuit is configured in a push-pull type as shown in Figure 1. The ON current value of the output current waveform shown in Figure 2 is a constant current and the OFF value set by RE is set by a time constant. Each output current can be calculated from the following expression:

- ON current = $1.5/R_E$ (A) (Output current setting pin voltage: $V_E \approx 1.5$ V)
- The OFF current time constant is proportional to the value of C_{B} .

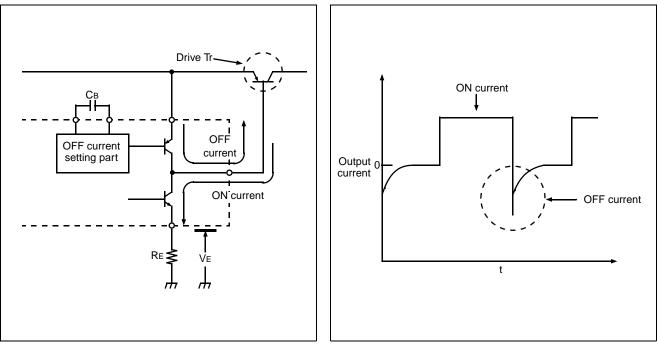


Fig.1 Output Circuit Diagram

200 nS

57

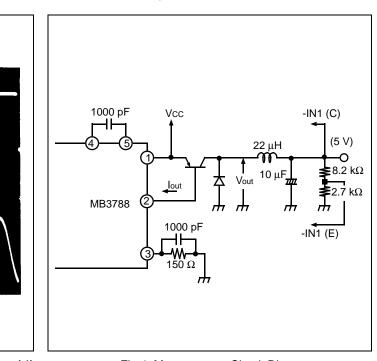


Fig.2 Output Current Waveform

Fig.4 Measurement Circuit Diagram

■ HOW TO SET TIME CONSTANT FOR TIMER & LATCH-TYPE SHORT-CIRCUIT PROTECTION CIRCUIT

If the load conditions of the switching regulator are stable, the outputs of comparators 1 and 2 do not change, so the SP comparator outputs a High level. At this time, the SCP pin (pin 15) is held at about 50 mV.

If the load conditions change suddenly due to a load short-circuit, for example, the output voltage of the comparator of the channel becomes a High-level signal (more than 2.1 V). Then, the SVP comparator outputs a Low level and transistor Q1 is turned off. The short-circuit protection capacitor C_{PE} externally connected to the SCP pin starts to charge.

 $V_{PE} = 50 \text{ mV} + t_{PE} \times 10^{-6}/C_{PE}$

 $0.65 = 50 \text{ mV} + t_{\text{PE}} \times 10^{-6}/C_{\text{PE}}$

 $C_{PE} = t_{PE} / 0.6 (s)$

Once the capacitor C_{PE} is charged to about 0.65 V, the SR latch is set and the output drive transistor is turned off. At this time, the duty cycle is made low and the output voltage of the SCP pin (pin 15) is held at Low level. This closes the SR latch input to discharge C_{PE} .

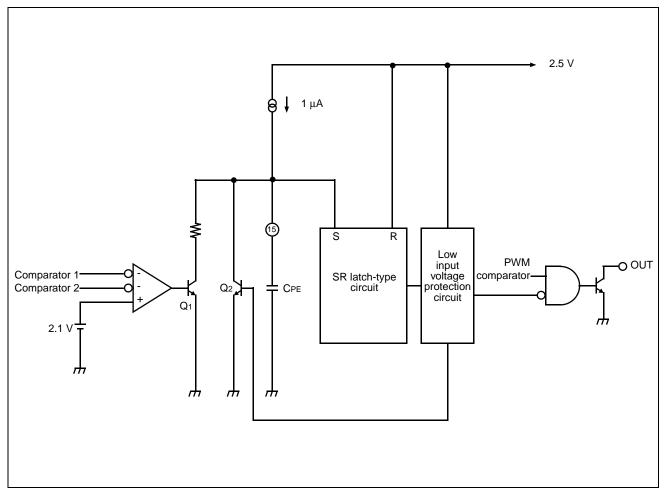


Fig. 5 Latch-Type Short-Circuit Protection Circuit

PROCESSING WITHOUT USING SCP PIN

If the timer and latch-type short-circuit protection circuit is not used, connect the SCP pin (pin 15) to GND as close as possible. Also, connect the input pin of each channel comparator to the Vcc pin (pin 11).

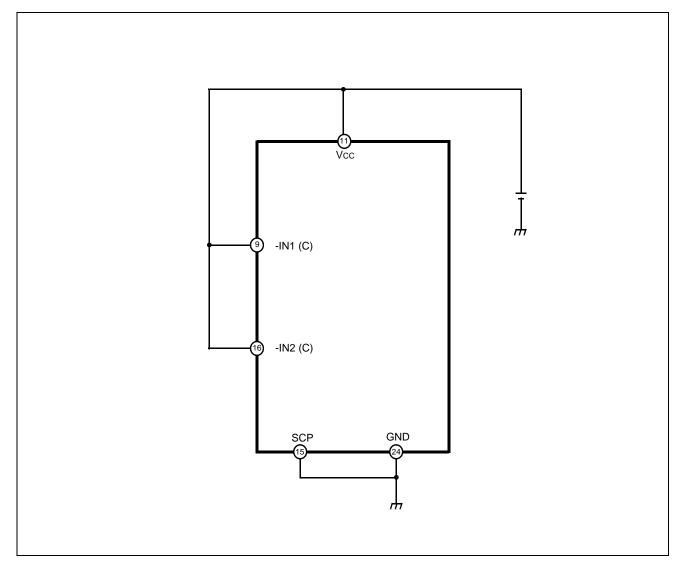


Fig. 6 Processing without using SCP Pin

■ EQUIVALENT SERIES RESISTANCE OF SMOOTHING CAPACITOR AND STABILITY OF DC/DC CONVERTER

The equivalent series resistance (ESR) of the smoothing capacity in a DC/DC converter has a great effect on the loop phase characteristics.

The ESR causes a small delay at the capacitor with a series resistance of 0 (Figures 8 and 9), thus improving system stability. On the other hand, using a smoothing capacitor with a low ESR reduces system stability. Therefore, attention should be paid to using semiconductor electrolytic capacitors (such as OS capacitors) or tantalum capacitors with a low ESP. (Phase margin reduction by using an OS capacitor is explained on the next page.)

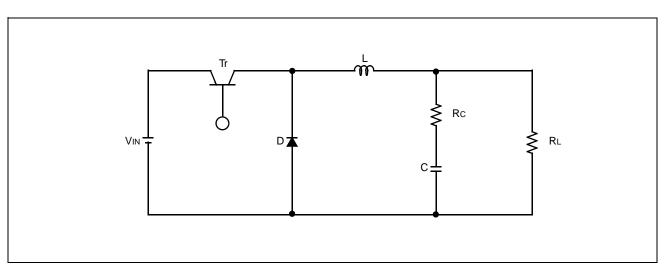


Fig. 7 Basic Voltage-Drop Type DC/DC Converter Circuit

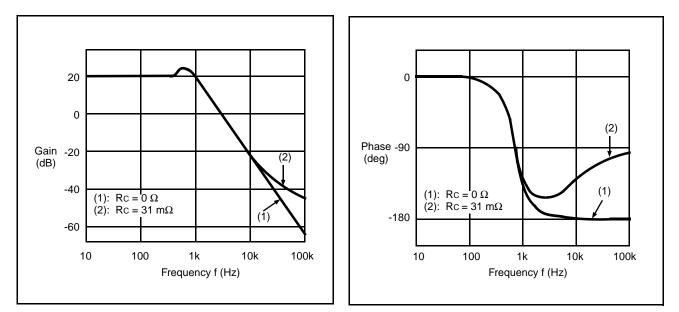
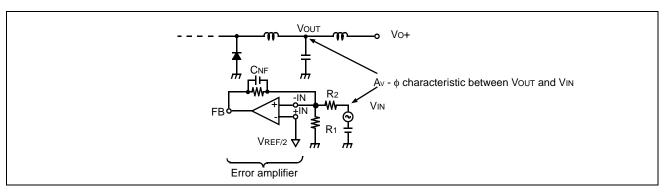


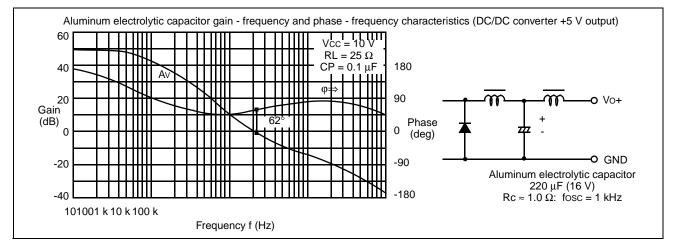
Fig.8 Gain - Frequency Characteristic

Fig.9 Phase - Frequency Charecteristic

(Reference Data)

The phase margin is halved by changing the smoothing capacitor from an aluminum electrolytic capacitor (Rc = 1.0Ω) to a semiconductor electrolytic capacitor (OS capacitor: Rc = 0.2Ω) with a low ESR (Figures 11 and 12).







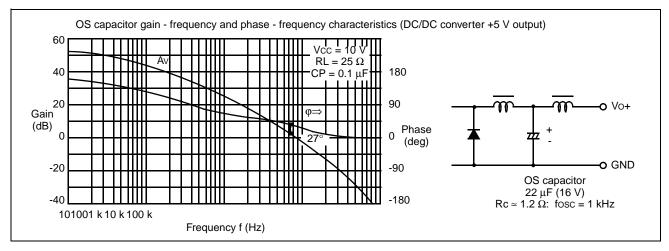
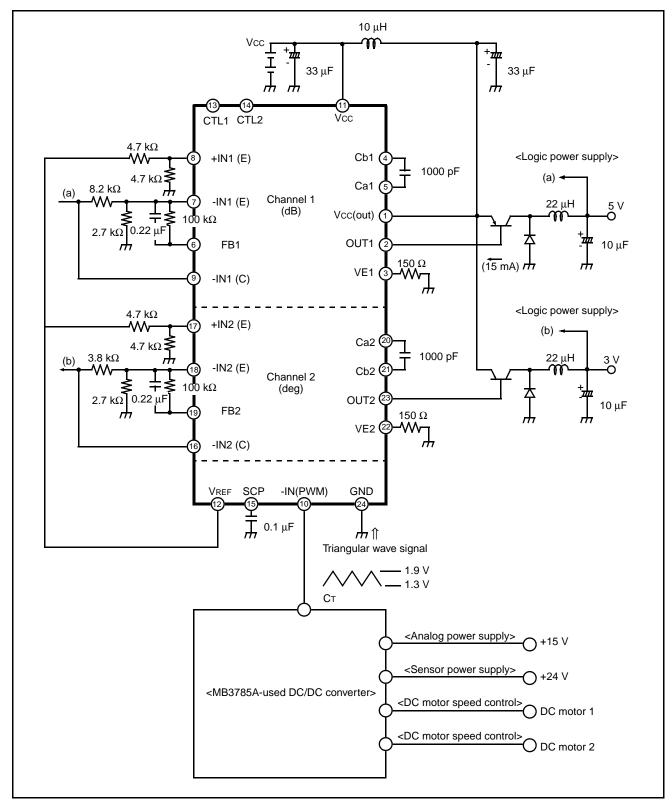


Fig.12 Phase - Frequency Characteristic Curves

■ APPLICATION CIRCUIT



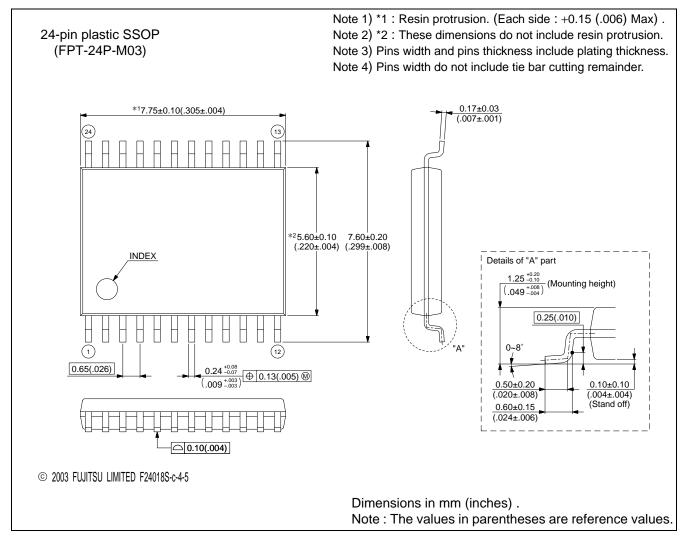
NOTES ON USE

- Take account of common impedance when designing the earth line on a printed wiring board.
- Take measures against static electricity.
 - For semiconductors, use antistatic or conductive containers.
 - When storing or carrying a printed circuit board after chip mounting, put it in a conductive bag or container.
 - The work table, tools and measuring instruments must be grounded.
 - The worker must put on a grounding device containing 250 k Ω to 1 M Ω resistors in series.
- Do not apply a negative voltage
 - Applying a negative voltage of –0.3 V or less to an LSI may generate a parasitic transistor, resulting in malfunction.

ORDERING INFORMATION

Part number	Package	Remarks
MB3788PFV	24-pin Plastic SSOP (FPT-24P-M03)	





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